

**IGBT Modules** 

### IGBT Module (X series) 600V / 20A / IPM

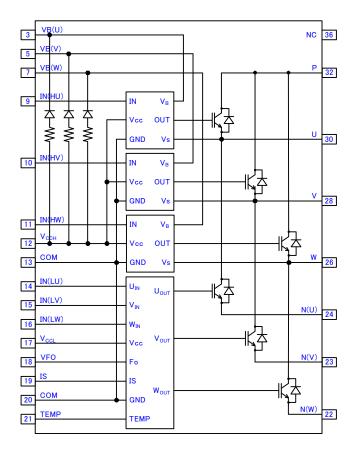
#### ■ Features

Low-side IGBTs are separate emitter type Short circuit protection Temperature sensor output function Under voltage protection Fault signal output function Input interface : TTL(3.3V/5V)Active high logic

### ■ Applications

AC 100 ~ 240V three phase inverter drive for small power AC motor drives (such as compressor motor drive for air conditioner, compressor motor drive for heat pump applications, fan motor drive, ventilator motor drive)

### ■ Terminal assign and Internal circuit





Pin No.	Pin Name	Pin Description
3	VB(U)	High-side bias voltage for
		U-phase IGBT driving
5	VB(V)	High-side bias voltage for
		V-phase IGBT driving
7	VB(W)	High-side bias voltage for
		W-phase IGBT driving
9	IN(HU)	Signal input for high side U-phase
10	IN(HV)	Signal input for high side V-phase
11	IN(HW)	Signal input for high side W-phase
12	V <sub>CCH</sub>	High-side control supply
13	COM	Common supply ground
14	IN(LU)	Signal input for low side U-phase
15	IN(LV)	Signal input for low side V-phase
16	IN(LW)	Signal input for low side W-phase
17	V <sub>CCL</sub>	Low-side control supply
18	VFO	Fault output
19	IS	Over current sensing voltage input
20	COM	Common supply ground
21	TEMP	Temperature sensor output
22	N(W)	Negative bus voltage input for
		W-phase
23	N(V)	Negative bus voltage input for
		V-phase
24	N(U)	Negative bus voltage input for
		U-phase
26	W	Motor W-phase output
28	V	Motor V-phase output
30	U	Motor U-phase output
32	Р	Positive bus voltage input
36	NC	No Connection

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### ■ Absolute Maximum Ratings (T<sub>vj</sub>=25°C, T<sub>c</sub>=25°C, V<sub>cc</sub>=15V unless otherwise specified)

	Items	Symbol	Conditions	Unit	Remarks
	DC bus voltage	$V_{DC}$	450	V	Note*1
	Bus voltage (surge)	$V_{\mathrm{DC(surge)}}$	500	V	Note*1
	Collector-Emitter voltage	V <sub>CE(chip)</sub>	600	V	V <sub>IN</sub> =0V
	Collector current	I <sub>C</sub>	20	Α	Note*2
<u></u>	Peak collector current	,	40	Α	V <sub>CC</sub> ≧15V,V <sub>B(*)</sub> ≧15V Note*2,*3,*4
nverter block	t dan concolor current	I <sub>CP</sub>	30	Α	V <sub>CC</sub> ≧13V,V <sub>B(*)</sub> ≧13V Note*2,*3,*4
blo	Forward current	I <sub>F</sub>	20	Α	Note*2
웃	Peak diode forward current	I <sub>FP</sub>	40	Α	Note*2
	Collector power dissipation	$P_{D\_IGBT}$	41.0	W	per single IGBT T <sub>C</sub> =25°C
	FWD power dissipation	$P_{D_FWD}$	27.8	W	per single FWD T <sub>C</sub> =25°C
	Virtual junction temperature	$T_{vj}$	150	°C	
	Operating virtual junction temperature		-40~+150	00	Note*8
	(under switching conditions)	$T_{\text{vjop}}$		°C	
	High-side supply voltage	V <sub>CCH</sub>	-0.5~20	V	Applied between VCCH-COM
	Low-side supply voltage	V <sub>CCL</sub>	-0.5~20	V	Applied between VCCL-COM
		$V_{\text{VB(U)-COM}}$			Applied between
	High-side bias absolute voltage	V <sub>VB(V)-COM</sub>		V	VB(U)-COM,VB(V)-COM,
		$V_{VB(W)\text{-COM}}$			VB(W)-COM
Contro	High-side bias voltage for IGBT gate driving	V <sub>B(U)</sub> V <sub>B(V)</sub> V <sub>B(W)</sub>	-0.5~20	V	Note*4
Control circuit block	High-side bias offset voltage	V <sub>U</sub> V <sub>V</sub> V <sub>W</sub>	-5~600	V	Applied between U-COM,V-COM,W-COM Note*5
ock	Input signal voltage	V <sub>IN</sub>	$-0.5 \sim V_{\text{CCH}} + 0.5$ $-0.5 \sim V_{\text{CCL}} + 0.5$	V	Note*6
	Input signal current	I <sub>IN</sub>	3	mA	sink current
	Fault signal voltage	$V_{FO}$	-0.5~V <sub>CCL</sub> +0.5	V	Applied between VFO-COM
	Fault signal current	I <sub>FO</sub>	1	mA	sink current
	Over current sensing input voltage	$V_{IS}$	-0.5~V <sub>CCL</sub> +0.5	V	Applied between IS-COM
	Virtual junction temperature	$T_{vj}$	150	°C	
Ор	erating case temperature	$T_{C}$	-40 <b>~</b> +125	°C	See Fig.1-1
Sto	orage temperature	T <sub>stg</sub>	-40 <b>~</b> +125	°C	
Iso	lation voltage	V <sub>isol</sub>	AC1500	Vrms	Sine wave,60Hz t = 1min,Note*7

#### Note

- \*1 :  $V_{DC}$  is applied between P-N(U),P-N(V),P-N(W).
- \*2 : Pulse width and duty are limited by  $T_{vi}$ max.
- \*3 :  $V_{\rm CC}$  is applied between VCCH-COM, VCCL-COM.
- \*4 :  $V_{B(*)}$  is applied between VB(U)-U,VB(V)-V, VB(W)-W.
- \*5 : Over 13.0V applied between VB(U)-U,VB(V)-V, VB(W)-W. This IPM module might make incorrect response if the high-side bias offset voltage is less than -5V.
- \*6: Applied between IN(HU)-COM,IN(HV)-COM,IN(HW)-COM,IN(LU)-COM,IN(LV)-COM,IN(LW)-COM.
- \*7: Applied between shorted all terminal and IMS (Insulated Metal Substrate).
- \*8 : The maximum temperature during continuous operation is T<sub>vj</sub>=150°C. The operating conditions have to be decided so that the temperature is below T<sub>vj</sub>=150°C. Continuous operation at over T<sub>vj</sub>=150°C may result in degradation of product lifetime such as power cycling capability.



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#### **■** Electrical characteristics

### ●Inverter block (T<sub>vj</sub>=25°C unless otherwise specified)

Description	Symbol	Condit	ions	min.	typ.	max	Unit
Zara gata valtaga pollogtar gurrant	,	V <sub>CE</sub> =600V	T <sub>vi</sub> =25°C	-	-	1	mA
Zero gate voltage collector current	I <sub>CE</sub>	V <sub>IN</sub> =0V	T <sub>vj</sub> =125°C	-	-	10	mA
Collector-Emitter saturation voltage		V <sub>CC</sub> = +15V	1 <sub>C</sub> =2A		0.90	1.10	10
		V <sub>B(*)</sub> =+15V	<i>T</i> <sub>vj</sub> =25°C	-	0.90	1.10	
	V	V <sub>IN</sub> =5V	/ <sub>C</sub> =20A		1.60	1.90	V
	V <sub>CES(sat)</sub>	V <sub>IS</sub> =0V	<i>T</i> <sub>vj</sub> =25°C	-	1.00	1.90	V
		Note *3, *4	/ <sub>C</sub> =20A		1.75	2.10	
			<i>T</i> <sub>vj</sub> =125°C	-	1.73	2.10	
Forward voltage	V <sub>F</sub>	/ <sub>F</sub> =20A	<i>T</i> <sub>vj</sub> =25°C	-	1.70	2.05	V
Forword voltage	V F	V <sub>IN</sub> =0V	T <sub>vj</sub> =125°C	-	1.50	-	V
Turn-on time	t on	V <sub>DC</sub> = 300V		0.60	1.00	1.35	
Turn-on delay time	t d(on)	I <sub>C</sub> = 20A		-	0.80	-	
Turn-on rise time	t <sub>r</sub>	V <sub>CC</sub> =15V		-	0.20	-	
V <sub>CE-</sub> I <sub>C</sub> cross time of turn-on	t <sub>c(on)</sub>	V <sub>B(*)</sub> =15V		-	0.40	0.65	
Turn-off time	t off	T <sub>vj</sub> = 125°C		-	1.00	1.40	μS
Turn-off delay time	t d(off)	V <sub>IN</sub> =0V <-> 5V		-	0.90	-	
Turn-off fall time	t <sub>f</sub>	V <sub>IS</sub> =0V		-	0.10	-	
V <sub>CE-</sub> I <sub>C</sub> cross time of turn-off	t <sub>c(off)</sub>	See Fig.2-1		-	0.15	0.30	
Reverse recovery time	t rr	Note *1, *3, *4		-	0.20	-	



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### Control circuit block

( $T_{\rm vj}$ =25°C,  $V_{\rm CC}$ =15V,  $V_{\rm B(*)}$ =15V,  $V_{\rm IN}$ =0V,  $V_{\rm IS}$ =0V unless otherwise specified)

Description	Symbol	Condit	ions	min.	typ.	max	Unit
Circuit current of low-side	1	V <sub>CCL</sub> =15V	V <sub>IN</sub> =5V	-	0.6	0.9	mA
Circuit current of low-side	I <sub>CCL</sub>	V <sub>CCL</sub> =15V	V <sub>IN</sub> =0V	-	0.6	0.9	IIIA
Circuit current of high-side	1	V <sub>CCH</sub> =15V	V <sub>IN</sub> =5V	1	1.25	1.9	mA
	I <sub>CCH</sub>	V <sub>CCH</sub> =15V	V <sub>IN</sub> =0V	-	1.25	1.9	Ш
Circuit current of bootstrap circuit	,	V <sub>B(U)</sub> =15V	V <sub>IN</sub> =5V	-	-	0.2	Λ
(per one unit)	/ <sub>CCHB</sub>	$V_{B(V)} = 15V$ $V_{B(W)} = 15V$	V <sub>IN</sub> =0V	•	-	0.2	mA
Input signal threshold voltage	$V_{\text{th(on)}}$			1	2.1	2.6	V
input signal tilleshold voltage		$V_{\text{th(off)}}$ Note*9 $V_{\text{th(hys)}}$ PW $\geq$ 0.7 $\mu$ s		8.0	1.3	1	V
Input signal threshold				0.35	0.8	_	V
hysteresis voltage	th(hys)			0.55	0.0	-	V
Operational input pulse	<i>t</i>	V <sub>IN</sub> =0V to 5V ris	se up	0.5			0
width of turn-on	t <sub>IN(ON)</sub>	Note*6,*9		0.5	-	-	μS
Operational input pulse	4	V <sub>IN</sub> =5V to 0V fa	ll down	0.7			
width of turn-off	t <sub>IN(OFF)</sub>	Note*6,*9		0.7	-	-	μS
Input current	I <sub>IN</sub>	V <sub>IN</sub> =5V Note*6		0.7	1.0	1.5	mA
Input pull-down resistance	R <sub>IN</sub>	Note*6		3.3	5.0	7.2	kΩ
•	V	V <sub>IS</sub> =0V,V <sub>FO</sub> terr	minal pull up	4.9			V
Fault output voltage	$V_{FO(H)}$	to 5V by $10k\Omega$		4.8			V
	$V_{FO(L)}$	V <sub>IS</sub> =1V,I <sub>FO</sub> =1m	A	-	-	0.95	V
Fault output pulse width	t <sub>FO</sub>	Note*10 See Fi	g.2-2, 2-3	20	-	-	μS



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#### **■** Electrical characteristics

#### Control circuit block (continued)

Description	Symbol	Conditions	min.	typ.	max	Unit
Over current protection voltage level	V <sub>IS(ref)</sub>	V <sub>cc</sub> =15V Note*3,11	0.455	0.48	0.505	V
Over current protection delay time	$t_{d(IS)}$	See Fig.2-2	0.3	8.0	1.3	μS
Output voltage of		Note*12 $T_{vj(LVIC)}=90^{\circ}C$	2.63	2.77	2.91	V
temperature sensor	$V_{\text{(temp)}}$	T <sub>vj(LVIC)</sub> =25°C	0.88	1.13	1.39	V
LVIC protection	ToH	Note*12	136	143	150	ç
T <sub>OH</sub> hysteresis	T <sub>OH(hys)</sub>	See Fig.2-6	4	10	20	C
V <sub>CC</sub> under voltage trip level of low-side	V <sub>CCL(OFF)</sub>		10.3	-	12.5	V
V <sub>CC</sub> under voltage reset level of	1/	<i>T</i> <sub>vj</sub> <150℃	10.0		12.0	V
low-side	V <sub>CCL(ON)</sub>	See Fig.2-3	10.8	-	13.0	V
V <sub>CC</sub> under voltage hysteresis	V <sub>CCL(hys)</sub>	1	-	0.5	-	V
V <sub>CC</sub> under voltage trip level of high-side	V <sub>CCH(OFF)</sub>		8.3	-	10.3	V
V <sub>CC</sub> under voltage reset level of	1/	T <sub>vi</sub> <150℃	0.0		10.0	V
high-side	V <sub>CCH(ON)</sub>	See Fig.2-4	8.8	-	10.8	V
V <sub>CC</sub> under voltage hysteresis	V <sub>CCH(hys)</sub>	1	-	0.5	-	V
V <sub>B</sub> under voltage trip level	V <sub>B(OFF)</sub>	T <sub>vj</sub> <150°C	10.0	-	12.0	V
V <sub>B</sub> under voltage reset level	V <sub>B(ON)</sub>	See Fig.2-5	10.5	-	12.5	V
V <sub>B</sub> under voltage hysteresis	V <sub>B(hys)</sub>		-	0.5	-	V
	V <sub>F(BSD)</sub>	$T_{vj}$ =25°C $I_{F(BSD)}$ =10mA	0.9	1.4	1.9	V
Forward voltage of bootstrap diode	V <sub>F(BSD)</sub>	$T_{vj}$ =25°C $I_{F(BSD)}$ =100mA	2.3	4.3	6.3	V

#### Note

<sup>\*9 :</sup> This IPM module might make incorrect response if the input signal pulse width is less than  $t_{\rm IN(on)}$  and  $t_{\rm IN(off)}$  .

<sup>\*10:</sup> Fault signal is asserted corresponding to "Over-current protection", "Under-voltage protection" at low-side, and "Overheat protection".

Under the condition of "Over-current protection", "Under-voltage protection" or "Overheat protection", the fault signal is asserted continuously while these conditions are continuing. However, the minimum fault output pulse width is minimum 20µsec even if very short failure condition (which is less than 20µs) is triggered.

<sup>\*11:</sup> Over current protection is functioning only for the low-side arms.

<sup>\*12:</sup> Fig.1-1 shows the measurement position of temperature sensor.



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#### ■ Thermal characteristic( $T_c$ =25°C)

Description	Symbol	min.	typ.	max	Unit
Junction to case thermal resistance (per single IGBT) Note*13	R <sub>th(j-c)_IGBT</sub>	-	ı	3.05	°C/W
Junction to case thermal resistance (per single FWD) Note*13	R <sub>th(j-c)_FWD</sub>	-	-	4.5	°C/W

Note\*13: Thermal compound with good thermal conductivity should be applied evenly with about

#### ■ Mechanical characteristics( $T_c$ =25°C)

Description	Symbol	Conditions	min.	typ.	max	Unit
Mounting torque of screws	M <sub>S</sub>	Mounting screw : M3	0.59	0.69	0.98	Nm
		The AL-IMS part:	-50		100	
Heat-sink side flatness		See (A1),(A2) of Fig.1-2 and Fig.1-3	-30	_		μm
	-	The resin case part:	-200	-	0	
		See (B1),(B2) of Fig.1-2 and Fig.1-3				
Weight	-	-	-	9.3	-	g
		Solder temp : 260 ±5°C				
Resistance to soldering heat	-	Immersion time : 10±1s	-	-	1	time
		Solder alloy : Sn-Ag-Cu type				

<sup>+100</sup>μm~+200μm on the contactingsurface of this device and heat-sink.

# Innovating Energy Technology

### 6MBP20XSF060-50

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### ■ Recommend operation conditions(Note\*17)

Description	Symbol	min.	typ.	max	Unit
DC bus voltage	V <sub>DC</sub>	0	300.0	400.0	V
High-side bias voltage for IGBT gate driving	V <sub>B(*)</sub>	13.0	15.0	18.5	V
High-side supply voltage	V <sub>CCH</sub>	13.5	15.0	16.5	V
Low-side supply voltage	V <sub>CCL</sub>	13.5	15.0	16.5	V
Control supply variation (under swiching conditions)	ΔV <sub>B</sub>	-1	-	1	V/μs
Control supply variation (under swiching conditions)	ΔV <sub>CC</sub>	-1	-	1	V/μS
Input signal voltage	V <sub>IN</sub>	0	-	5	V
Voltage for current sensing	V <sub>IS</sub>	0	-	5	V
Potential difference of between COM to N (including surge)	V <sub>COM_N</sub>	-5	-	5	V
Dead time for preventing arm-short (T <sub>C</sub> ≤125°C)	t <sub>DEAD</sub>	1.0	-	-	μS
Output current (Note*14)	Io	-	-	16.0	A rms
Minimum input pulse widht (Note*15 Note*16)	PW <sub>IN(on)</sub>	0.5	-	-	μS
Minimum input pulse widht (Note*15,Note*16)	PW <sub>IN(off)</sub>	0.7	-	-	μS
PWM input frequency	$f_{PWM}$	-	-	20	kHz
Operating virtual junction temperature	$T_{\text{vjop}}$	-30	-	150	°C

#### Note

To reduce the loss of IGBT, please enlarge the pulse width more than the switching time of IGBT.

<sup>\*14 :</sup>  $V_{\rm DC}$  = 300V,  $V_{\rm CCH}$  =  $V_{\rm CCL}$  =  $V_{\rm B(")}$  = 15V, PF=0.8, Sinusoidal PWM, 3phase modulation,  $T_{\rm vj}$  ≤ 150°C ,  $T_{\rm c}$  ≤ 100°C ,  $f_{\rm PWM}$  = 5kHz,  $f_{\rm O}$  = 200Hz, Ks=0.9

<sup>\*15 :</sup>In the pulse width of 0.5us, the loss of IGBT increases for the saturation operation.

<sup>\*16 :</sup>This IPM module might response according to input signal pulse even when the input signal pulse width is less than  $PW_{IN(on)}$  and  $PW_{IN(on)}$ .

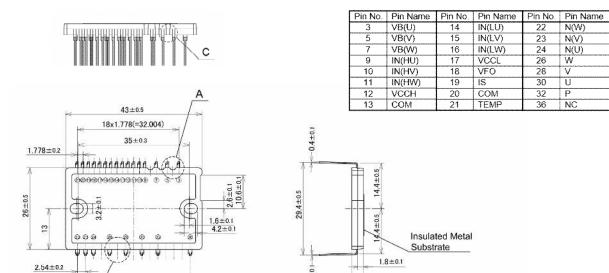
<sup>\*17:</sup> Recommended operating conditions are conditions for guaranteeing that the product operates normally. If it is used beyond this condition, operation and reliability may be adversely affected.

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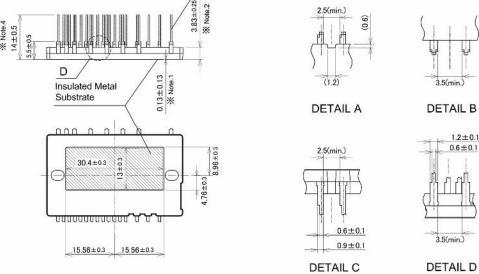
#### ■ Packing outline dimensions ( $T_c=25^{\circ}C$ )

4x2.54 (=35.56)

В



3.7±0.1



Solder Plating

※Note.1

IMS(Insulated Metal Substrate) is deliberately protruded to improve the thermal conductivity between IMS and heat-sink.

≫Note 2

The thickness from the package surface to the back side includes the IMS.

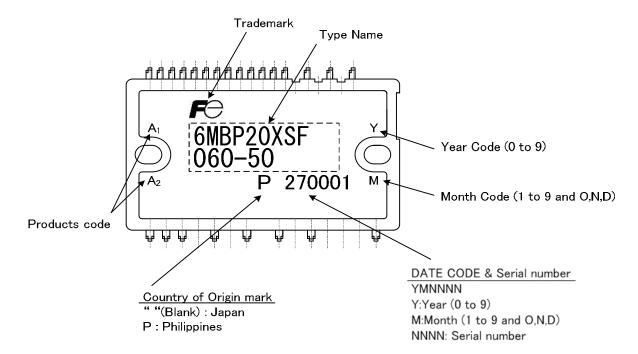
Thickness of the case part of the package outer wall. (excluding the IMS and marking surface)  $\times$  Note.4

Height of the terminal and height of the stopper part including IMS.

Unit: mm

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#### ■ Marking



#### Note

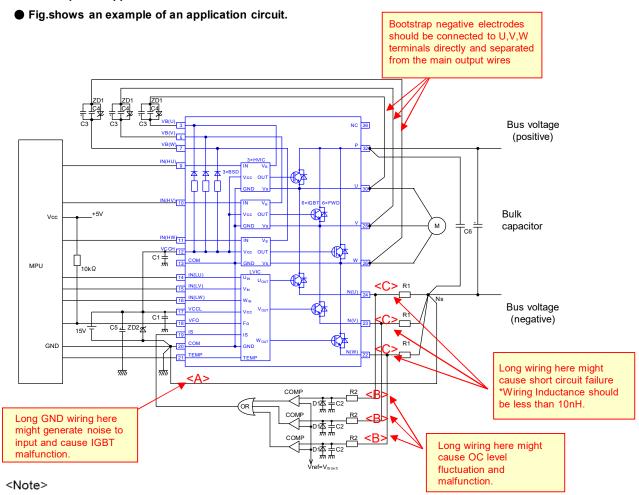
Product code A<sub>1</sub> means current ratings , and "M" is marked.

Product code  $A_2$  means variations , and "F" is marked.



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An example of application circuit.



- Input signal for drive is High-Active. There is a pull-down resistor built in the IC input circuit. To prevent malfunction, the
  wiring of each input should be as short as possible. When using R-C coupling circuit, make sure the input signal level
  meet the turn-on and turn-off threshold voltage.
- 2. By the function of the HVIC, it is possible of the direct coupling to microprocessor (MPU) without any photo-coupler or pulse-transformer isolation.
- 3. VFO output is open drain type. It should be pulled up to the positive side of a 5V power supply by a resistor of about  $10k\Omega$ .
- 4. To prevent erroneous protection, the wiring of (A), (B), (C) should be as short as possible.
- 5. The time constant R2-C2 of the protection circuit should be selected approximately 1.5µs.

  Over current (OC) shutdown time might vary due to the wiring pattern. Tight tolerance, temp-compensated type is recommended for R2, C2.
- 6. Please set the threshold voltage of the comparator reference input to be same as the IPM OC trip reference voltage VIS(ref).
- 7. Please use high speed type comparator and logic IC to detect OC condition quickly.
- 8. If negative voltage of R1 at the switching timing is applied, the schottky barrier diode D1 is recommended to be inserted parallel to R1.
- 9. All capacitors should be mounted as close to the terminals of the IPM as possible. (C1, C4: narrow temperature drift, higher frequency and DC bias characteristic ceramic type are recommended, and C3, C5: narrow temperature drift, higher frequency and electrolytic type.)
- 10. To prevent surge destruction, the wiring between the snubber capacitor and the P terminal, Ns node should be as short as possible. Generally a 0.1μ to 0.22μF snubber capacitor (C6) between the P terminal and Ns node is recommended.
- 11. Two COM terminals (13 & 20 pin) are connected inside the IPM, it must be connected either one to the signal GND outside and leave another one open.
- It is recommended to insert a zener-diode (22V) between each pair of control supply terminals to prevent surge destruction.
- 13. If signal GND is connected to power GND by broad pattern, it may cause malfunction by power GND fluctuation. It is recommended to connect signal GND and power GND at only a point.

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Fig.1-1: The measurement position of temperature sensor.

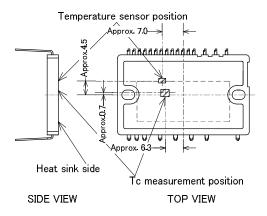


Fig.1-2: The measurement position of heat sink flatness.

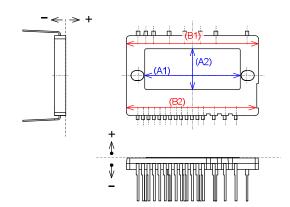
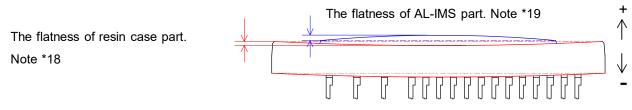


Fig.1-3:

The magnified cross section image of warp direction.

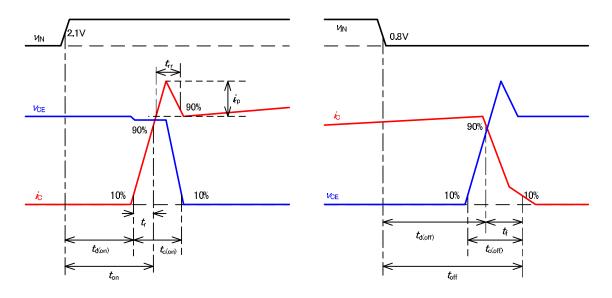
- \* This image is a stretched drawing (Not true scale)
- \* A positive value means the AL-IMS direction. A negative value means the marking surface direction.



#### Note

- \*18: The virtual datum level assumes a straight line to link both ends of the resin case.
- \*19: The virtual datum level assumes a straight line to link both ends of the AL-IMS.

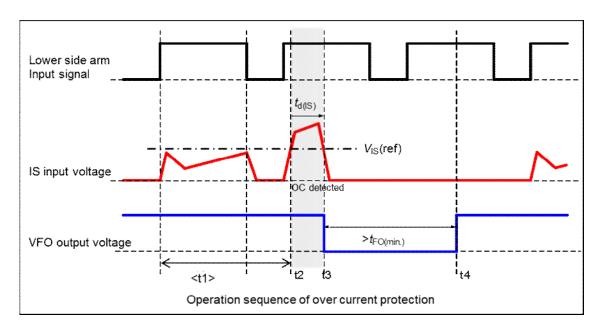
Fig.2-1 Switching waveforms



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Fig.2-2\_Operation\_sequence\_of over\_current\_protection



- <t1>: IS input voltage does not exceed  $V_{\text{IS(ref)}}$ , while the collector current of the lower side IGBT is under the normal operation.
- t2 : When IS input voltage exceeds  $V_{\rm IS(ref)}$ , the OC is detected.
- t3 : The fault output VFO is activated and all lower side IGBT shut down simultaneously after the over current protection delay time  $t_{d(IS)}$ . Inherently there is dead time of LVIC in  $t_{d(IS)}$ .
- t4: After the fault output pulse width  $t_{\rm FO}$ , the OC is reset. Then next input signal is activated.

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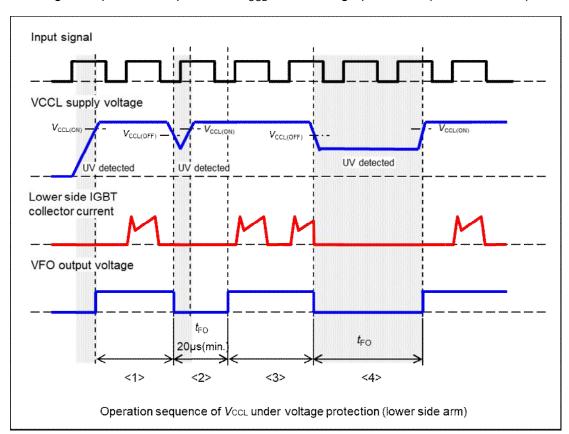


Fig.2-3 Operation sequence of  $V_{CCL}$  under voltage protection (lower side arm)

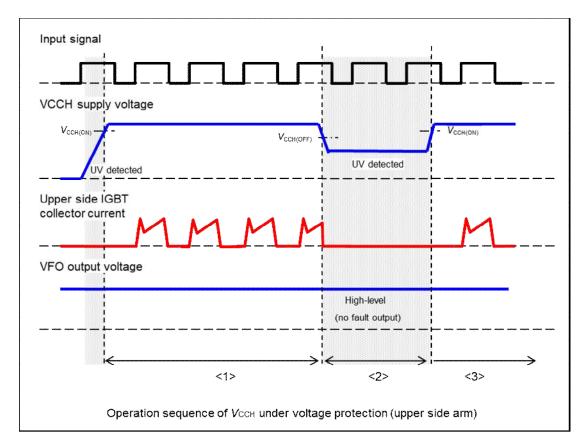
When VccL is under 4V, UV and fault output are not activated.

- <1> When  $V_{\text{CCL}}$  is under  $V_{\text{CCL}(\text{ON})}$ , all lower side IGBTs are off state. After  $V_{\text{CCL}}$  rises to  $V_{\text{CCL}(\text{ON})}$ , the fault output VFO is released (high level). And the LVIC starts to operate, then next input is activated.
- <2> The fault output VFO is activated when VccL falls below VccL(OFF), and all lower side IGBT remains off state.

  When the voltage drop time is less than 20µs, the fault output pulse width is generated minimum 20µs and all lower side IGBTs are off state in spite of input signal condition during that time.
- <3> UV is reset after tFo when VCCL exceeds VCCL(ON) and the fault output VFO is reset simultaneously.
  - And the LVIC starts to operate, then next input is activated.
- <4> When the voltage drop time is more than too, the fault output pulse width is generated and all lower side IGBTs are off state in spite of input signal condition during the same time.

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Fig.2-4 Operation sequence of  $V_{\rm CCH}$  under voltage protection (upper side arm)



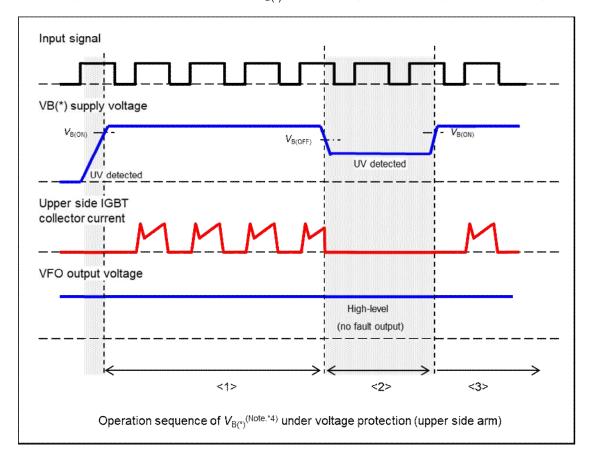
- <1> When Vccн is under Vccн(ом), the upper side IGBT is off state.

  After Vccн exceeds Vccн(ом), the HVIC starts to operate. Then next input is activated.

  The fault output VFO is constant (high level) not depending on Vccн.
- <2> After  $V_{\rm CCH}$  falls below  $V_{\rm CCH(OFF)}$ , the upper side IGBT remains off state. But the fault output VFO remains at high level.
- <3> The HVIC starts to operate after UV is reset, then next input is activated.

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Fig.2-5 Operation sequence of  $V_{B(*)}$  under voltage protection (upper side arm)

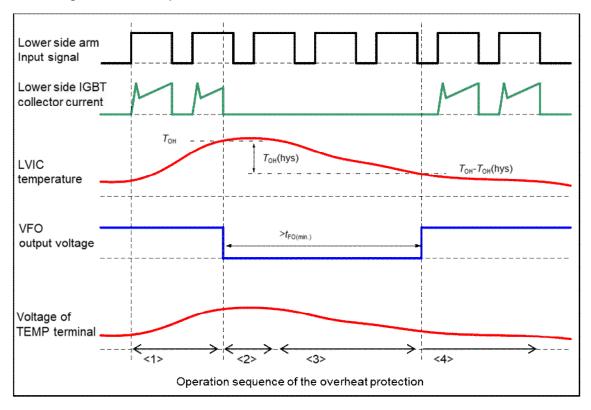


- <1> When  $V_{B(U)}$ ,  $V_{B(V)}$  or  $V_{B(W)}$  are under  $V_{B(ON)}$ , the corresponding upper side IGBTs are off state. After  $V_{B(U)}$ ,  $V_{B(V)}$  or  $V_{B(W)}$  exceed  $V_{B(ON)}$ , the corresponding upper side IGBTs start to operate. Then next input is activated. The fault output VFO is constant (high level) not depending on  $V_{B(Y)}$ . (Note\*20)
- <2> After  $V_{B(U)}$ ,  $V_{B(V)}$  or  $V_{B(W)}$  fall below  $V_{B(OFF)}$ , the corresponding upper side IGBTs remain off state. But the fault output VFO keeps high level.
- <3> The HVIC starts to operate after UV is reset, then next input is activated.

Note \*20: The fault output is not given HVIC bias conditions.

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Fig.2-6 Overheat protection



This function is applied to "6MBP\*\*XSF060-50".

The IPM has overheat protection (OH) function by monitoring the LVIC temperature.

The  $T_{\rm OH}$  sensor position is shown in Fig.1-1.

- <1> The collector current of the lower side IGBT is under the normal operation while the LVIC temperature does not exceed  $T_{\rm OH}$ .
- <2> The IPM shutdown all lower side IGBTs while the LVIC temperature exceeds  $T_{\mathrm{OH}}$ .
- <3> The TEMP terminal continue to output the voltage which correspond to temperature of LVIC even if IPM is in OH condition.
- <4> The fault status is reset when the LVIC temperature drops below ( $T_{\rm OH}$ - $T_{\rm OH(hys.)}$ ). All lower side IGBTs restart to normal operation.

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